

# THE ALGORITHM OF TUNING CONTROLLERS TO THE MODEL OBJECT WITH INERTIA, ADVANCE AND TIME DELAY IN THE CASCADE CONTROL SYSTEM

Irina Cojuhari

Technical University of Moldova

**Abstract** – This paper proposes a tuning algorithm of linear controllers P, PI, PID in the multiple-loop feedback control systems. The control object consists of two subprocesses, which are described by the dynamical models with advance, inertia (first and third order) and time delay. The controllers in the internal contour and in the external contour tuning use the maximal stability degree method. P and PI controllers are used in the internal contour and P, PI, PID controllers are used in the external contour.

**Keywords** –multiple-loop feedback control system, internal contour, external contour, time delay, maximal stability degree method.

## 1. INTRODUCTION

Many tuning methods of typical controllers are used at the projecting of multiple-loop control systems: frequency method, criteria (of modulus) method etc [1,2,4]. The procedure of tuning controllers in the multiple-loop feedback control system becomes difficult. In this paper, it is proposed to use the maximal stability degree method (M.S.D) for tuning of typical controllers P, PI, PID for a class of control objects' models with inertia, advance and time delay, which are connected in cascade, represented by two subprocesses and, as result with two regulating loops.

## 2. THE ITERATIVE ALGORITHM OF TUNING CONTROLLERS

The multiple-loop feedback control system is represented by two contours: internal contour with controller's transfer function  $H_{R2}(s)$  and subprocess  $H_{F2}(s)$ , and external contour with controller's transfer function  $H_{R1}(s)$  and subprocess  $H_{F1}(s)$  (Fig. 1). It is recommended to carry out the tuning of controllers first in the internal contour then in the external contour.

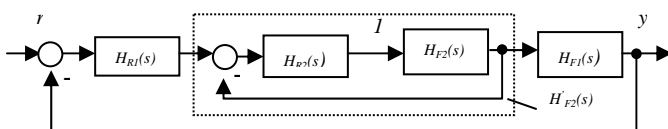


Fig.1 - The multiple-loop feedback control system

The control object consists of two inertial subprocesses with the transfer functions:

$$H_{F1}(s) = \frac{T_1s + 1}{(T_2s + 1)(T_3s + 1)(T_4s + 1)} = \frac{c_0s + c_1}{d_0s^3 + d_1s^2 + d_2s + d_3}, \quad (1)$$

where  $c_0 = T_1$ ;  $c_1 = 1$ ;  $d_0 = T_2T_3T_4$ ;

$d_1 = T_2T_3 + T_2T_4 + T_3T_4$ ;  $d_2 = T_2 + T_3 + T_4$ ;  $d_3 = 1$ .

$$H_{F2}(s) = \frac{ke^{-\tau s}}{T_5s + 1}, \text{ with } T_5 < T_2, T_3, T_4. \quad (2)$$

In expressions (1) and (2) we have the notations:  $k$  are transfer coefficients of the subprocesses;  $T_1, T_2, T_3, T_4, T_5$  are time constants of respective subprocesses,  $\tau$  - time delay.

## 3. THE TUNING CONTROLLERS IN THE INTERNAL CONTOUR

The tuning of controller with transfer function (t.f.)  $H_{R2}(s)$  from internal contour to the subprocess with t.f.  $H_{F2}(s)$  is implemented. We assume that P and PI controllers are used.

P controller is tuned to the object with transfer function (2), applied M.S.D. method and tuning parameters of controller are determined from relation:

$$k_{p2} = \frac{e^{-\tau J}}{k} (T_5J - 1). \quad (3)$$

In the relation (3)  $J$  is the maximal stability degree which is chosen from the following condition  $J > 0$ .

To determinate the t.f. of the internal contour in case of tuning P and PI controllers it is proposed to approximate the value  $e^{-\tau s}$  with Pade approximant [2]:

$$e^{-\tau s} = \frac{1}{\tau s + 1}. \quad (4)$$

The t.f. of internal contour with P controller tuning is:

$$H'_{F2}(s) = \frac{H_{R2}(s) \cdot H_{F2}(s)}{1 + H_{R2}(s) \cdot H_{F2}(s)} = \frac{k_p k}{(T_5s + 1) \cdot (\tau s + 1) + k_p k} = \frac{k'}{l_0s^2 + l_1s + l_2}, \quad (5)$$

$$\text{where } k' = \frac{k_{p2}k}{1 + k_{p2}k}; l_0 = \frac{\tau T_5}{1 + k_{p2}k}; l_1 = \frac{\tau + T_5}{1 + k_{p2}k}; l_2 = 1.$$

PI controller is tuned to the object with the transfer function (2), applied the M.S.D. method and tuning parameters of controller are determined from relations [3,4]

$$k_{p2} = \frac{e^{-\tau J}}{k} (-\tau T_5 J^2 + (\tau + 2T_5)J - 1); \quad (6)$$

$$k_{i2} = \frac{e^{-\tau J}}{k_2} J^2 (-\tau T_5 J + \tau + T_5). \quad (7)$$

We can obtain the values of parameters  $k_{p2}, k_{i2}$ , changing the  $J > 0$  value, for that the performances of control system are sated.

The t.f. of internal contour with PI controller tuning, using expression (4) is:

$$H'_{F2}(s) = \frac{H_{R2}(s)H_{F2}(s)}{1+H_{R2}(s)H_{F2}(s)} = \frac{k(k_{p2}s+k_{i2})}{s(T_5s+1) \cdot (\tau s+1) + k(k_{p2}s+k_{i2})} = \frac{l_0s+l_1}{r_0s^3+r_1s^2+r_2s+r_3}, \quad (8)$$

$$\text{where } l_0 = \frac{k_{p2}}{k_{i2}}; \quad l_1 = 1; \quad r_0 = \frac{\tau T_5}{k_{i2}k}; \quad r_1 = \frac{(\tau+T_5)}{k_{i2}k}; \\ r_2 = \frac{1+k_{p2}k_2}{k_{i2}k_2}; \quad r_3 = 1.$$

#### 4. THE TUNING CONTROLLERS IN THE EXTERNAL CONTOUR

The structure block scheme of external contour is represented in the Fig. 2 a, b.

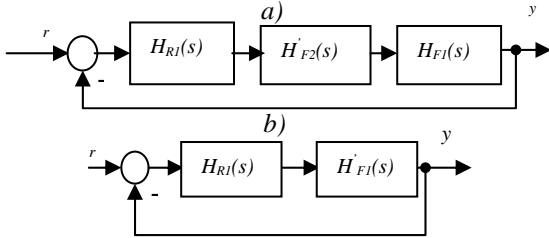


Fig. 2 - Structure block schema of external contour

For the tuning of P, PI, PID controllers in the external contour it is necessary to determine the equivalent transfer function of object (5) with P controller tuning in the internal contour with the t.f. of subprocess (1)

$$H'_{F1}(s) = H'_{F2}(s)H_{F1}(s) = \frac{k'}{(l_0s^2+l_1s+l_2) \cdot (d_0s^3+d_1s^2+d_2s+d_3)} = \frac{c_0s+c_1}{(l_0s^2+l_1s+l_2) \cdot (d_0s^3+d_1s^2+d_2s+d_3)} = \frac{b_0s+b_1}{a_0s^5+a_1s^4+a_2s^3+a_3s^2+a_4s+a_5}, \quad (9)$$

$$\text{where } b_0 = \frac{k_{p2}kT_1}{1+k_{p2}k_2}; \quad b_1 = \frac{k_{p2}k}{1+k_{p2}k_2}; \quad a_0 = \frac{\tau T_2T_3T_4T_5}{1+k_{p2}k}; \\ a_1 = \frac{(\tau+T_5)T_2T_3T_4}{1+k_{p2}k} + \frac{\tau T_5(T_2T_3+T_3T_4+T_2T_4)}{1+k_{p2}k}; \quad a_2 = T_2T_3T_4 + \\ + \frac{(\tau+T_5)(T_2T_3+T_3T_4+T_2T_4)}{1+k_{p2}k} + \frac{\tau T_5(T_2+T_3+T_4)}{1+k_{p2}k}; \quad a_3 = (T_2T_3+T_3T_4+T_2T_4) + \\ + \frac{(\tau+T_5)(T_2+T_3+T_4)}{1+k_{p2}k} + \frac{\tau T_5}{1+k_{p2}k}; \quad a_4 = T_2+T_3+T_4 + \frac{\tau+T_5}{1+k_{p2}k}; \quad a_5 = 1.$$

For object with t.f. (9) P, PI, PID controllers can be tuned by applying the M.S.D. method using the relation from [3,4].

Control system with P controller:

$$k_p = \frac{a_0J^5 - a_1J^4 + a_2J^3 - a_3J^2 + a_4J - a_5}{b_1 - b_0J}. \quad (10)$$

Control system with PI controller:

$$k_p = \frac{-d_0J^6 + d_1J^5 - d_2J^4 + d_3J^3 - d_4J^2 + d_5J - d_6}{(b_1 - b_0J)^2}, \quad (11)$$

where  $d_0 = 5a_0b_0$ ;  $d_1 = 6a_0b_1 + 4a_1b_0$ ;

$$d_2 = 5a_1b_1 + 3a_2b_0; \quad d_3 = 4a_2b_1 + 2a_3b_0; \quad (12)$$

$$d_4 = 3a_3b_1 + a_4b_0; \quad d_5 = 2a_4b_1; \quad d_6 = a_5b_1.$$

$$k_i = \frac{-a_0J^6 + a_1J^5 - a_2J^4 + a_3J^3 - a_4J^2 + a_5J + k_pJ}{b_1 - b_0J}. \quad (13)$$

Control system with PID controller:

$$k_u = \frac{d_0J^7 - d_1J^6 + d_2J^5 - d_3J^4 + d_4J^3 - d_5J^2 + d_6J - d_7}{2(b_1 - b_0J)^4}, \quad (14)$$

where  $d_0 = 5a_0b_0^3$ ;  $d_1 = 68a_0b_0^2b_1 + 12a_1b_0^3$ ;

$$d_2 = 78a_0b_0b_1^2 + 42a_1b_0^2b_1 + 6a_2b_0^3;$$

$$d_3 = 30a_0b_1^3 + 50a_1b_0b_1^2 + 22a_2b_0^2b_1 + 2a_3b_0^3;$$

$$d_4 = 20a_1b_1^3 + 24a_2b_0b_1^2 + 8a_3b_0^2b_1; \quad d_5 = 12a_2b_1^3 + 12a_3b_0b_1^2;$$

$$d_6 = 6a_3b_1^3 + 2a_4b_0b_1^2 - 2a_5b_0^2b_1; \quad d_7 = 2a_4b_1^3 - 2a_5b_0b_1^2;$$

$$k_p = \frac{-d_0J^6 + d_1J^5 - d_2J^4 + d_3J^3 - d_4J^2 + d_5J - d_6 + 2k_dJ}{(b_0 - b_1J)^2}, \quad (15)$$

where  $d_0, d_1, d_2, d_3, d_4, d_5, d_6$  are determined from relations (12).

$$k_i = \frac{-a_0J^6 + a_1J^5 - a_2J^4 + a_3J^3 - a_4J^2 + a_5J - k_dJ^2 + k_pJ}{b_1 - b_0J}. \quad (16)$$

For the tuning of P, PI, PID controllers in the external contour it is necessary to determine the equivalent t.f. of object (8) with PI controller in the internal contour with the t.f. of subprocess (1)

$$H'_{F1}(s) = H'_{F2}(s)H_{F1}(s) = \frac{l_0s+l_1}{(r_0s^3+r_1s^2+r_2s+r_3)} \cdot \frac{c_0s+c_1}{(d_0s^3+d_1s^2+d_2s+d_3)} = \frac{b_0s^2+b_1s+b_2}{a_0s^6+a_1s^5+a_2s^4+a_3s^3+a_4s^2+a_5s+a_6}, \quad (17)$$

$$\text{where } b_0 = \frac{k_{p2}T_1}{k_{i2}}; \quad b_1 = \frac{k_{p2}}{k_{i2}} + T_1; \quad b_2 = 1; \quad a_0 = \frac{\tau T_2T_3T_4T_5}{k_{i2}k};$$

$$a_1 = \frac{(\tau+T_5)T_2T_3T_4}{k_{i2}k} + \frac{\tau T_5(T_2T_3+T_2T_4+T_3T_4)}{k_{i2}k};$$

$$a_2 = \frac{(1+k_{p2}k_2)T_2T_3T_4}{k_{i2}k_2} + \frac{(\tau+T_5)(T_2T_3+T_2T_4+T_3T_4)}{k_{i2}k} + \frac{\tau T_5(T_2+T_3+T_4)}{k_{i2}k};$$

$$a_3 = T_2T_3T_4 + \frac{(1+k_{p2}k_2)(T_2T_3+T_2T_4+T_3T_4)}{k_{i2}k_2} + \frac{(\tau+T_5)(T_2+T_3+T_4)}{k_{i2}k} + \frac{\tau T_5}{k_{i2}k};$$

$$a_4 = T_2T_3+T_2T_4 + \frac{(1+k_{p2}k_2)(T_2+T_3+T_4)}{k_{i2}k_2} + \frac{(\tau+T_5)}{k_{i2}k};$$

$$a_5 = T_2+T_3+T_4 + \frac{1+k_{p2}k_2}{k_{i2}k_2}; \quad a_6 = 1.$$

For object with t.f. (17) P, PI, PID controllers can be tuned by applying the M.S.D. method using the relations from [3,4].

Control system with P controller

$$k_{p1} = \frac{-a_0J^6 + a_1J^5 - a_2J^4 + a_3J^3 - a_4J^2 + a_5J - a_6}{b_0J^2 - b_1J + b_2}. \quad (18)$$

Control system with PI controller

$$k_{p1} = \frac{-d_0J^8 + d_1J^7 - d_2J^6 + d_3J^5 - d_4J^4 + d_5J^3 - d_6J^2 + d_7J - d_8}{(b_0J^2 - b_1J + b_2)^2}, \quad (19)$$

where  $d_0 = 5a_0b_0$ ;  $d_1 = 6a_0b_1 + 4a_1b_0$ ;

$$d_2 = 5a_1b_1 + 3a_2b_0 + 7a_0b_2; \quad d_3 = 4a_2b_1 + 2a_3b_0 + 6a_1b_2; \quad (20)$$

$$d_4 = 3a_3b_1 + a_4b_0 + 5a_2b_2; \quad d_5 = 2a_4b_1 + 4a_3b_2;$$

$$d_6 = a_5b_1 + 3a_4b_2 - a_6b_0; \quad d_7 = 2a_5b_2; \quad d_8 = a_6b_2,$$

$$k_{i1} = \frac{a_0J^7 - a_1J^6 + a_2J^5 - a_3J^4 + a_4J^3 - a_5J^2 + a_6J}{b_0J^2 - b_1J + b_2} + k_pJ. \quad (21)$$

Control system with PID controller



To compare the obtained results the tuning of the typical controller using Ziegler-Nichols method was made. The Fig. 5 presents the transition processes of the multiple-loop feedback control system for external contour in the case of using Ziegler-Nichols method: *a)* – transition process in the external contour, with P controller tuning in the internal contour and P, PI, PID controllers tuning in the external contour; *b)* – transition process in the external contour with PI controller tuning in the internal contour and P, PI, PID controller tuning in the external contour. The obtained values of respectively controllers are presented in Table 2, where the number of rows correspond with number of curves from Fig. 5 *a, b*.

Table 2 – The values of tuning controllers by Ziegler – Nichols method

Nr. curve s	$k_{p2}=1.025$	$k_{p2}=0.92$ $k_{i2}=0.17$
1	$k_{p1}=9$	$k_{p1}=5.1$
2	$k_{p1}=8.1$ $k_{i1}=0.048$	$k_{p1}=4.59$ $k_{i1}=0.039$
3	$k_{p1}=13.60$ $k_{i1}=0.064$ $k_d=2.6$	$k_{p1}=7.711$ $k_{i1}=0.05$ $k_{d1}=3.2$

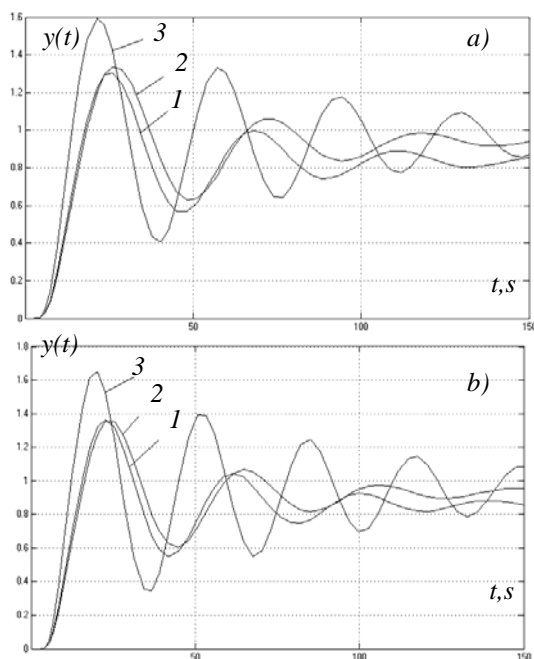


Fig. 5 - Transition processes of the multiple-loop feedback control system

The Fig. 6 *a)* presents the transition processes of the multiple-loop feedback control system, the Fig. 6 *b)* presents the repartition of the poles for the following cases: 1- control system with P controller tuning in the internal contour and PID in the external contour using the maximal stability degree method, 2 - control system with PI controller tuning in the internal contour and PID in the external contour using the maximal stability degree method, 3- control system with P controller tuning in the internal contour and PID in the external

contour using the Ziegler – Nichols method, 4- control system with PI controller tuning in the internal contour and PID in the external contour using the Ziegler – Nichols method.

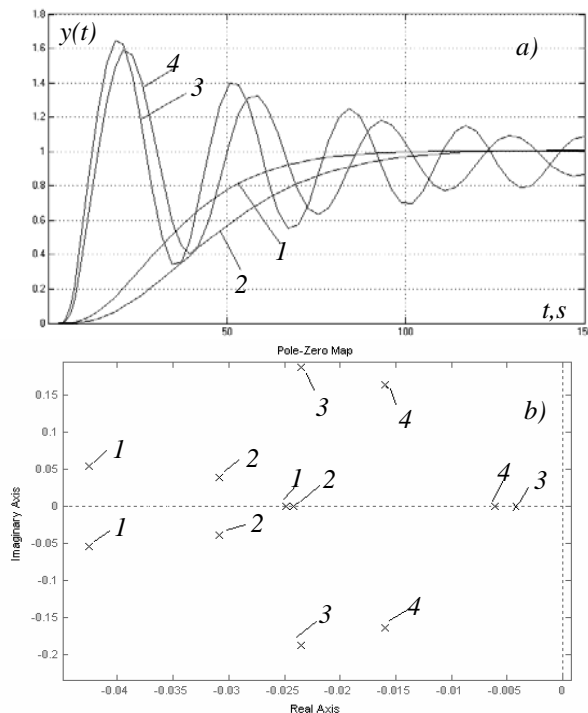


Fig. 6 - Transition processes of the multiple-loop feedback control system (*a*) and presentation of distribution of the systems' poles (*b*)

## 6. CONCLUSIONS

As a result, after tuning the P, PI, PID controller to the multiple-loop feedback control system with object's models (1), (2) with known parameters, the following conclusions can be made:

1. The tuning of P, PI controller in the internal contour in conformity with the maximal stability degree method permitted to obtain the high results varying the value of the  $J > 0$  and choosing the parameters of the respectively controllers for obtaining the sated performance of the internal contour.
2. The tuning of P, PI, PID controllers in the external contour using the maximal stability degree method permitted to obtain the high results varying the value of the  $J > 0$  and obtained the optimal value and the suboptimal values of the  $J$ , and choosing the set of the values of controllers' parameters for obtaining the sated performances.

## REFERENCES

- [1] I. Dumitrache și al. Automatizări electronice. - București: EDP, 1993, 660 p.
- [2] Ș. Preitl, R. E. Precup. Introducere în ingineria reglării automate, Timișoara: Editura Politehnică, 2001, 334 p.
- [3] B. Izvoareanu, I. Fiodorov, F. Izvoareanu. The Tuning of Regulator for Advance Delay Objects According to the Maximal Stability Degree Method / In: Proceedings of the 11th International Conference on Control Systems and Computer Science (CSCS-11), București, 1997, V.1., pp. 179-184.
- [4] Izvoareanu B., Fiodorov I., Cojuhari I. Tuning of Controllers to the Third Order Advance Delay Objects / In: Proceedings of the 5th

International Conference on Microelectronics and Computer Science (ICMCS-2007), Chişinău, 2007, V.I., pp. 250-253.

- [5] Tan N., Atherton D. P. Design of stabilizing PI and PID controllers. In: International Journal of Systems Science, 6/20/2006, Vol. 37. Issue 8, pp.543-554.